

# M2 Integration Circuits Systems

EA2 - Année Scolaire 2017-2018 (S1)

## Exam Durée 1h30 - Authorized documents and calculator

## Exercises

Exercise 1: From standard DCS1800 to receiver specifications Exercise 2: ADC spectral testing

Tous les exercices sont indépendants.

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### **Exercise 1: From standard DCS1800 to receiver specifications**

Frequency bands for DCS1800 standard are the following:

- ✓ uplink (mobile towards base station) : 1710 1785 MHz
- ✓ downlink (base station towards mobile) : 1805 1880 MHz

The inter channel spacing is 200 kHz. Main specifications are given in Table 1.

| Modulation type             | GMSK   |
|-----------------------------|--|
| Transmit band               | 1710 – 1785 MHz                                  |
| Receive band                | 1805 – 1880 MHz                                  |
| Channel bandwidth           | 200 KHz  |
| Reference sensitivity level | -102 dBm   |
| BER                         | $10^{-3} \rightarrow \text{SNRmin} = 7\text{dB}$ |

Table I: Main specifications of DCS1800 standard

From these data, determine:

Q1.1 – The thermal noise power, at ambient temperature (T=300K) in the channel at the receiving antenna. (Boltzmann constant  $k = 1,38 \ 10^{-23} \ J/K$ )

Q2.2 – The maximal Noise Figure (NF) allowed for the receiver.



Oscillateur Local

The above schematic shows the receiver architecture. At the antenna, the highest interferer level in the useful band is -26dBm. The technology used for the LNA design limits the LNA maximum output power to -10dBm and leads to a LNA NF of 6 dB.

Q2.3 – To avoid saturation of the LNA, what is its gain maximum value? We assume that the gain is set to this value for the following.

Q2.4 – Assuming RF filter is not degrading NF and after the mixer, the NF degradation is negligible, calculate the maximal NF allowed for the mixer.

#### **Exercise 2: ADC spectral testing**

Shown below is a 4096 point FFT of the output of a 9-bit A/D converter for full scale sinusoidal input.



Q2.1 – What is the SFDR of the ADC?

The FFT plot only shows fundamental and  $3^{rd}$  harmonic tones. It is therefore plausible to assume that main source of nonlinearity is  $3^{rd}$  order distortion:

$$v_{out} = a_1 v_{in} + a_3 v_{in}^3$$

Infinite number of ADC bits is assumed in above transfer characteristic for simple calculation. With a single-tone input,

$$v_{in} = 1/2 V_{FS} \sin(\omega t)$$

Given that the amplitude of the 3<sup>rd</sup> harmonic is 42dB below that of the fundamental

Q2.2 – Express  $a_3$  in function of  $a_1$  and V<sub>FS</sub>.

You should use the following formula:  $sin3a = 3sina - 4sin^3a$ 

And in order to simplify, we will suppose that  $a_1 V_{FS}/2 >>3 a_3 V_{FS}^3/32$ 

Q2.3 – Calculate the end points of the out/in transfer function ( $v_{out}$  at - $V_{FS}/2$  and at  $V_{FS}/2$ )

Q2.4 – Calculate the FS linear gain value and conclude that the linearized transfer function is given by the following equation:

$$v'_{out} = 1.032 a_1 v_{in}$$

Q2.5 – Compute the  $|INL|_{max}$  of the converter in LSBs.

Note that: the definition of INL requires that the offset and gain errors of the ADC are corrected before ie offset = 0 and linear gain = 1. Then find the expression of  $v_{outcor}$  before computing  $|INL|_{max}$ 

$$v_{outcor} (-V_{FS}/2) = -V_{FS}/2, v_{outcor} (0) = 0, v_{outcor} (V_{FS}/2) = V_{FS}/2$$

For ease of computation, you can assume the transfer function has infinite number of steps.

Q2.6 –Use the rule of thumb expression for SFDR versus INL :

$$SFDR = 20 \log \left(\frac{2^B}{INL_{/LSB}}\right)$$
 where B is the number of bits of the ADC.

Assuming INL = 3.1LSB, calculate an approximate value for SFDR, compare the result with the more accurate one you found in Q2.1.

Q2.7 - Calculate the total quantization noise amplitude of a 9-bit ADC in dBFS

Q2.8 – Given 4096 FFT points, quantization noise floor on the spectrum plot should be 33.11 dB lower than quantization noise amplitude ( $10\log 10(4096/2) = 33.11$ ). Since this is the spectrum for a real ADC with both quantization noise and circuit generated noise (assume that the actual FFT shows noise floor of about -80dBFS), the difference is then contributed by circuit noise, which can be assumed uncorrelated with quantization noise.

Compute the approximate noise contribution by the circuit in dBFS

Q2.9 – Find the ENOB for this ADC based on SNR only.

Q2.10 – Find the ENOB for this ADC based on SNDR.